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Optical Switching Applications of Delayed
Feedback Nonlinear Systems

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I. Planning and System Design

The goal of this project was to develop an optical switching system based on an optical nonlinear delayed-feedback system previously developed at ATR by T. Aida and P. Davis. A secondary task was optimizing the feedback system for the switching application. Nonlinear delayed-feedback systems are useful because for a relatively simple implementation they exhibit a wide range of complex functions, such as data storage, random data generation, and adaptive data generation.

The primary project was to set up an optical switching system using the output of the nonlinear fiber-delayed-feedback system. The switching technique that was chosen is direct injection of an optical pump signal into a lasing device. Optical switching has previously been demonstrated in traveling-wave or Fabry-Perot optical amplifiers using a separate CW laser as the probe source. However, direct injection is a simpler technique combining the optical source and wavelength conversion mechanism into a single device.

The lasers used in this experiment are conventional single-stripe long-wavelength lasers which are available commercially. Lower switching thresholds have been demonstrated using a multisection laser with a saturable absorber to quench lasing in the absence of a pump signal. The saturable absorber also induced bistability, which is an interesting subject for further investigation. Bistable devices were not readily available for use here. Distributed feedback (DFB) lasers were used to provide single-frequency output.

The price of pigtailed lasers is about an order of magnitude higher than for chip-mounted or can-mounted devices. The cost of building a general-purpose fiber pigtailed setup is comparable to buying a pigtailed laser. However, a general-purpose setup can be used to compare switching in many different kinds of lasers at relatively low cost. For example, Fabry-Perot lasers vs DFB lasers, and gain-depletion pumping vs gain-enhancement pumping.

A major component of the project was construction of a 1.3 μm semiconductor optical amplifier. A commercially available Erbium-doped fiber amplifier was used to get high-power signals for switching at 1.55 μm . A semiconductor amplifier can be used to directly do optical switching in the amplifier itself due to gain saturation or four-wave mixing. A semiconductor amplifier was constructed for backup as a 1.3 μm amplification source as well as a frequency-conversion component.

II. Nonlinear System Optimization

This section discusses optimization of the nonlinear feedback system for repetition rate and output power. A high pulse repetition rate (greater than 100MB/s) is important in order to demonstrate the relevance of this work to current optical switching problems. Repetition rate problems that were addressed are loop delay and frequency response flatness.

High output power is desirable to minimize the amount of amplification needed for optical switching. High power is particularly important because this type of system has a fairly small contrast in power between the '1' (high power) state and '0' (low power) state.

Two nonlinear feedback systems had been built at ATR prior to the start of this research project. The first is a low frequency demonstration version which has a maximum clock frequency of 25.5 MHz and uses a 1.3 μ m laser. The second is a high performance system with a theoretical maximum clock frequency around 1 GHz that uses a 1.55 μ m laser. This switching work focused on the second system in order to achieve high switching rates.

Nonlinear feedback systems require flat frequency response from the fundamental oscillation frequency (1/4 times the loop delay) to about 3-5 times the highest desired oscillation frequency in order to allow the harmonics associated with fast transitions. A very small amplitude imbalance <0.5dB has been shown by Aida and Davis (JQE 1992) to cause preferential selection of particular bit patterns.

The frequency response of the optical modulator/detector response is shown in figure 2. The resonance around 20 MHz is associated with the acoustic resonant of the modulator, which is coupled into the optical performance by the piezoelectric effect. The nonlinear system had a strong tendency to lock into this resonance.

The first problem to be solved is the effect of this resonance. The resonance itself is not easily eliminated. It is important to push the total nonlinear loop delay time above this 20 MHz resonance by minimizing the amplifier delay and electrical and optical path delays. Minimizing the loop delay also has the important benefit of increasing the repetition rate for a given bit length pattern, allowing for faster switching demonstrations. This delay was minimized in collaboration with P. Davis and T. Aida by shortening the electrical and optical path lengths.

When building very broadband microwave amplifiers, small resonances unavoidable occur the frequency response. Cascading amplifiers for higher gain also increases the amplitude of gain ripple, frequently to unacceptable levels. The amplifier chain should be composed of amplifiers each with a different design to avoid cascading resonances. Ideally the amplifiers would even be from different manufacturers. However, one of the goals is to include DC response, and there are few sources of high-frequency DC-coupled sources.

The bandwidth of the components is sufficient to allow GHz oscillation. However, due to the frequency ripple, equalization of the frequency response is essential in order to realize this GHz potential. Figure 3 shows the initial design of an equalizer, and a general equalizer design which can be built and then customized for various ripple functions. A design was optimized and constructed by Nogawa. Using this equalizer, GHz oscillation was achieved. In order to limit the number of data bits in the system, coaxial low-pass filters were purchased from Mini-Circuits. By using different bandwidths, the number of bits in the system was varied from 3 to 21.

The optical modulator had about 3dB higher loss than originally measured at the factory. With more optical power and reduced loss in the optical modulator, it should be possible to get oscillation with only two amplifiers, reducing the delay and ripple of the system.

Further work to be done on optimizing the system are:

- (a) Reduce insertion loss of the modulator
- (b) Shorter fiber leads on the modulator
- (c) Further equalizer optimization
- (d) Use a higher-power laser

III. Experimental Results

A key task in this switching project was getting good coupling from the laser into the fiber. Both a Hoya A85 aspheric lens and an NSG gradient-index lens were used with good results. It was not possible to measure the absolute coupling efficiency, as a broad area photodetector would be needed to capture all of the light. The NSG lens has a very high numerical aperture, and should capture most of the light emitted from the laser. Both lenses were able to get a coupling efficiency of 3dB compared to the light collimated by the NSG lens.

A 1.3 semiconductor optical amplifier was designed and assembled using an NTT Electronics amplifier chip. The switching demonstration ended up being done at 1.55 μ m, and the amplifier was not used in the experiment.

Several switching options were investigated. The first was optical pumping. In this approach the output of the nonlinear system would be in the absorption band of another laser. When the nonlinear system signal pump was on, it would generate carrier in the probed laser which would decrease the threshold and increase the output power. This is a noninverting switching operation. This was attempted with the pump laser at 1.3 μ m and the probe laser at 1.55 μ m. However, very little modulation of the probe laser, with 1mW of pump power producing a few tens of uW in the probe laser. Apparently the absorption of the 1.55 μ m probe lasers is not sufficiently high at 1.3 μ m.

The second approach was polarization switching. In this approach, the laser has a desired polarization (usually TE). The opposite polarization (usually TM) is injected, causing the laser to switch polarization. A polarizer following the laser converts the polarization switching to amplitude modulation. This was also attempted, but although a variety of lasers were tested, all of them exhibited a strong preference for TE polarization, and would not switch.

The final switching method that was tested was injection locking. In this method a signal is injected which is at a different wavelength from the lasers natural frequency, causing the lasers wavelength to change. An optical filter can be used to convert the wavelength change to amplitude modulation. This switching technique was tested with a number of lasers at both 1.3 μ m and 1.55 μ m, and was universally successful.

Finally, the optical system was used as a source of control signals for driving a WDM signal switching system. A novel implementation shown in Figure 4 was developed which exploits the repetitive nature of the optical system output in order to obtain complementary switched signals from identical switching systems. A technique for implementing this approach was developed by splitting the pump pulse and combining the two output pulses using a polarization coupler. This configuration is optimum for polarization switching. It can also be used for injection locking, and has the useful property that the two lasers are combined with orthogonal polarizations, which is helpful for separating them without an optical filter.

DFB lasers were used as the probe devices, since WDM (wavelength division multiplexing) implies that the output is at a particular wavelength, while Fabry-Perot lasers have an output at many wavelengths. The final configuration used an NEC laser at 1545.*nm and a Mitsubishi laser at 1549.*nm. The pump laser was a Fujitsu laser at 1552.*nm. Using different manufacturers was very helpful in getting different wavelengths, as lasers ordered from one source tended to be at the same wavelength, particularly if ordered at the same time.

An optical filter was implemented by collimating the light from the fiber, reflecting it off of a grating, and coupling the light back into another fiber. This filter was used to separate the pump and probe signals for time-domain analysis. This filter had a bandwidth of a few nm, and a loss of more than 6dB. The reflectivity of the grating was 60%. The origin of this high loss needs further investigation, but it could be due to either the long path length or imperfections in the grating. The fiber to fiber spacing was about 4cm. Long distance between the gradient index fiber couplers leads to increased loss as the beam front quality tends to degrade with distance. The grating surface quality appeared to be poor, which can also lead to increased coupling loss due to beam front degradation.

The switching characteristics were examined using an HP8112A pulse generator into the optical modulator. The rise and fall time of the outputs was approximately equal to the rise and fall time from the modulator, which was on the order of 2ns. This indicates that the switching time is $<2\text{ns}$.

Switching was done using several different bit patterns which were selected from the nonlinear system by varying the microwave low-pass filter which limits the number of bits in the system. Figure 5a shows the output from the NEC and Mitsubishi lasers using the 933MHz filter. Complementary output patterns occur due to the 3m fiber delay from the coupler to the NEC laser. The intensity modulated pattern directly from the nonlinear system is transformed to on/off pulse information by the thresholding effect of the switching. Figure 5b shows the output of the Mitsubishi laser using the 1830MHz filter. From this data, optical switching time of the lasers is $<1\text{ns}$.

IV. Conclusion

The goal of this project was to use the nonlinear feedback system for WDM/routing optical switching. The nonlinear system was optimized for this purpose, and the output was used to switch two lasers at different wavelengths with a complementary bit pattern.

V. Acknowledgments

We would like to thank T. Watanabe for his support and encouragement. We would like to thank T. Aida (NHK) for his assistance in optimizing the nonlinear system hardware for optical switching. We would like to thank T. Miyazaki, K. Inagaki, H. Kawamura, and H. Shimotaira for lending advice and equipment, and M. Hosoda for his expertise with oscilloscopes and endless supply of miscellaneous parts.

Appendix 1: Mechanical design

Mechanical parts were designed for a number of purposes related to optical switching and for a semiconductor optical amplifier. This is a description of general mechanical parts that were designed, and their purpose.

<i>Part</i>	<i>Purpose</i>
Aspheric Lens Holder	permanently mount laser-fiber focusing lenses
Aspheric Lens Tube	mount collimating or focusing lenses
Base Modification	make Fiber Base and Mounting Base from commercial part
Beam Collimator Holder	hold fiber collimators on stand or in mirror mount
Chip Carrier Mount	attach chip-mounted lasers to TE cooler
DIN Connector Bracket	attach DIN connector to Mounting Plate
Dual Aspheric Lens Holder	mount collimating lens and hold lens tube
FC Holder Stand Mount	hold FC fiber optic connector for coupling
FC Rotation Base	hold FC Rotation Plate
FC Rotation Plate	hold and rotate polarization preserving fiber
Fiber Base	hold Fiber Rotation Plate on 5-axis positioner
Fiber Strain Relief	prevent fiber movement from degrading coupling
Grating Mount	mount and protect grating
GRIN Lens Holder	mount GRIN SELFOC lenses
Lens Arm	attach various lens holders to XYZ positioner
Lens Arm Rest	protect lenses in box
Mounting Base	attach Mounting Plate to XYZ positioner fixed plate
Mounting Plate	hold laser/TE cooler and DIN Connector Bracket
Pigtailed Laser Plate	hold butterfly laser packages, fits standard cover
Submarine Can Mount	hold submarine laser can packages
Thor 6mm Lens Mount	hold Thor Labs aspheric lens mount

This is a description of semiconductor amplifier parts. The semiconductor amplifier also uses some of the general parts listed above.

<i>Part</i>	<i>Purpose</i>
Amplifier Lens Arm	attach Dual Aspheric Lens Mount to XYZ positioner
Aspheric Tube Clamp	clamp Isolator Assembly to Semiconductor Amp Stand
Cable Clamp	hold wires on amplifier assembly
DIN Angle Bracket	attach DIN connector to amplifier assembly
Dual Aspheric Lens Mount	hold Isolator Assembly near amplifier
Isolator Assembly	hold isolator and lens tubes
Semiconductor Amp Base	hold Semiconductor Amp Stand
Semiconductor Amp Stand	hold semiconductor amplifier
TE spacer	use in place of thermoelectric cooler

Appendix 2: Related Literature

The following is a reference list of papers related to this work. Many of these papers have optical bistability as the main topic, but also discuss optical switching.

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Fig. 1 Schematic diagram of the optical nonlinear delayed-feedback system. The equalizer is inserted between amplifiers A1 and A2. Amplifier A3 may not be necessary if the output light levels are reduced (ie. loss in optical arm is reduced).

Fig. 2. Nonlinear system component gain spectrum (S21) (a) modulator+photodetector detector (b) single amplifier

Fig. 3. Gain Equalizer (a) Structural design (b) Gain spectrum of "+3.5dB at 1.5GHz" equalizer designed for equalization of circuit with modulator+photodetector+two amplifiers. (Injection Loss IL =3.9dB at < 10MHz, 0.4dB at 1.5GHz; VSWR<1.40)

Fig. 4. Experimental configuration for optical switching experiment

Fig. 5. Example of TDM/WDM switching at 699MHz (a) Outputs from (a) LD 1 laser (NEC 1545nm DFB LD) and (b) LD 2 laser (Mitsubishi DFB 1548nm DFB LD) with (c) nonlinear system output, showing conversion of the amplitude modulated nonlinear system signal to WDM pulses

Fig. 6. Example of TDM/WDM switching at 699MHz (a) Output of LD 2 and (b) output of nonlinear system, corresponding to bit sequence "..1011011..."

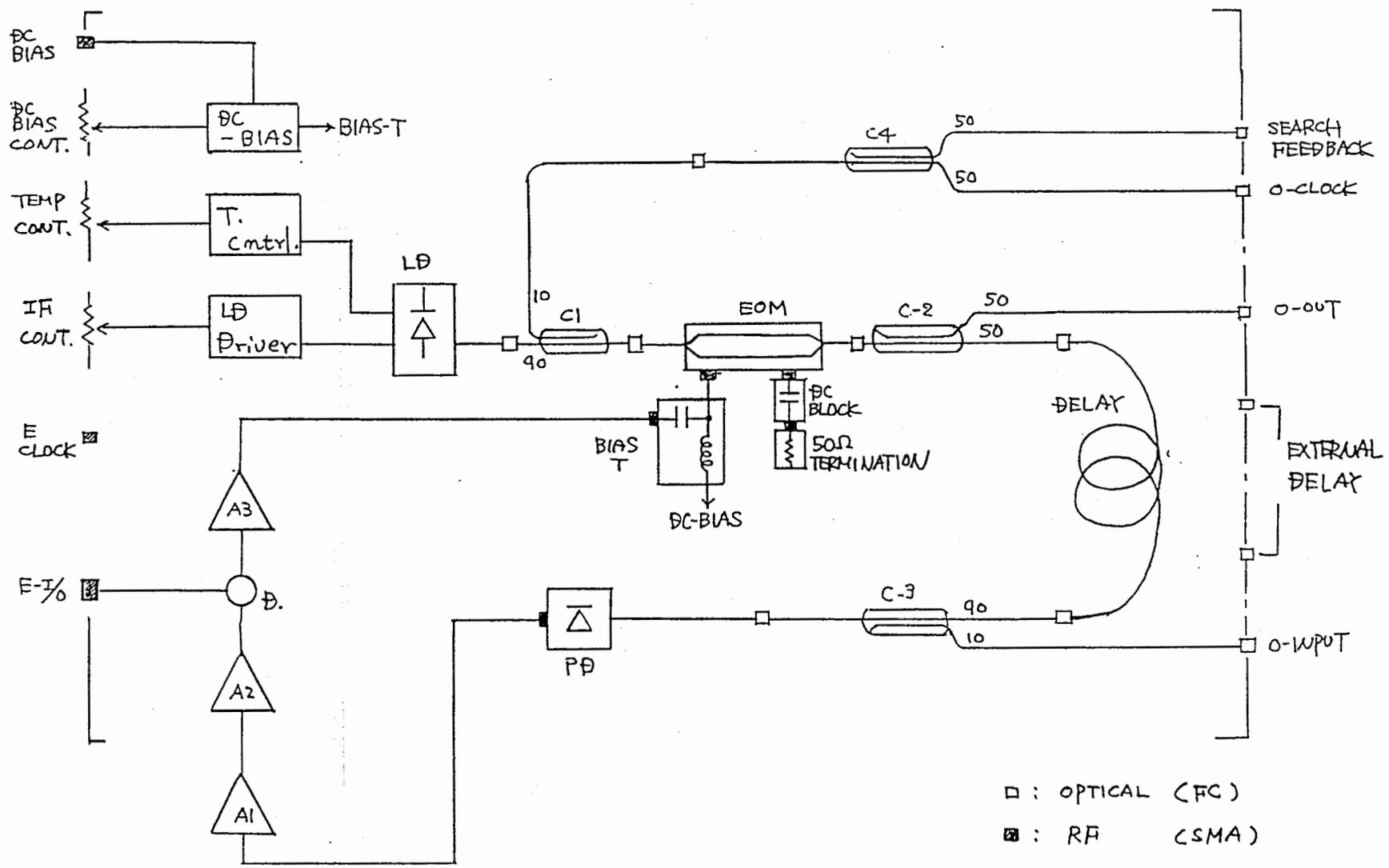
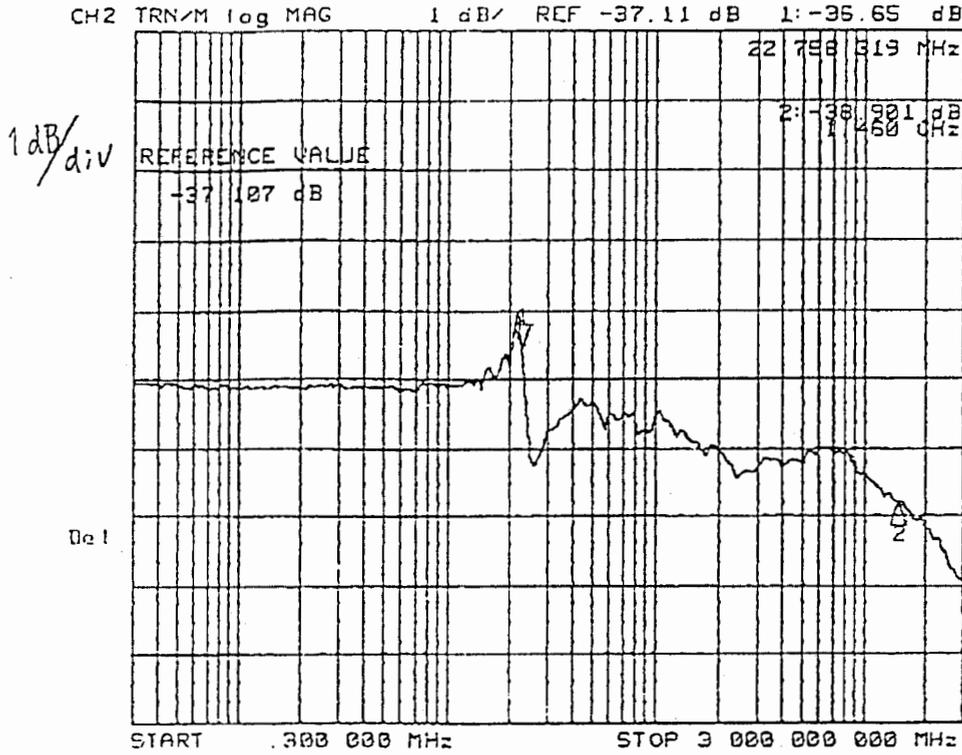
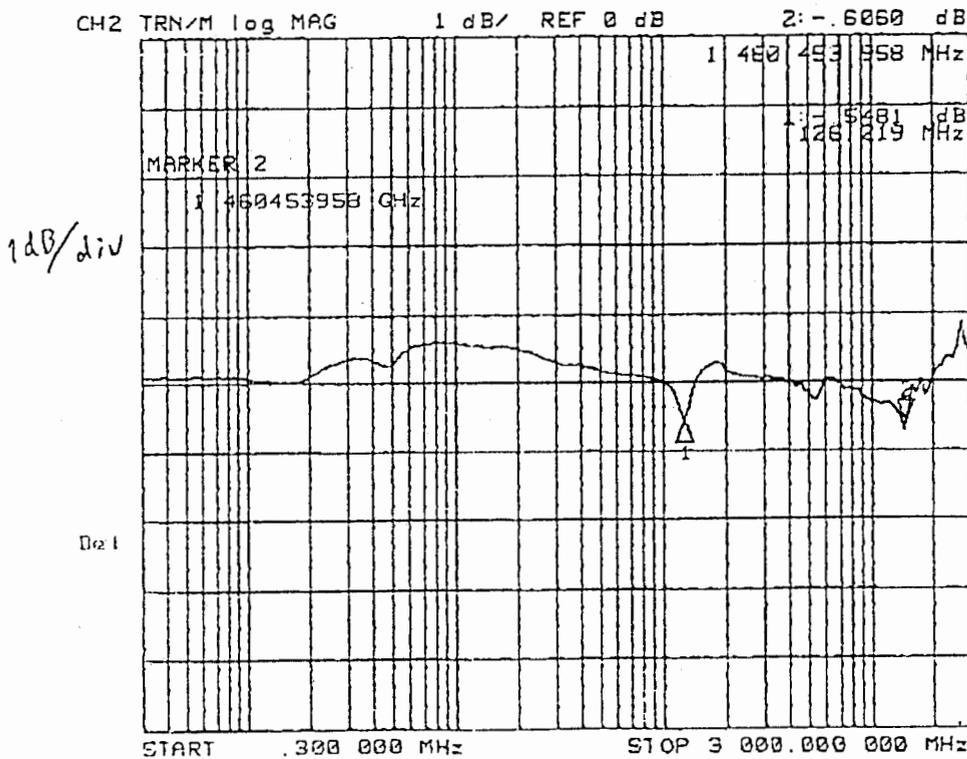


Fig. 1

Nonlinear System Component Response

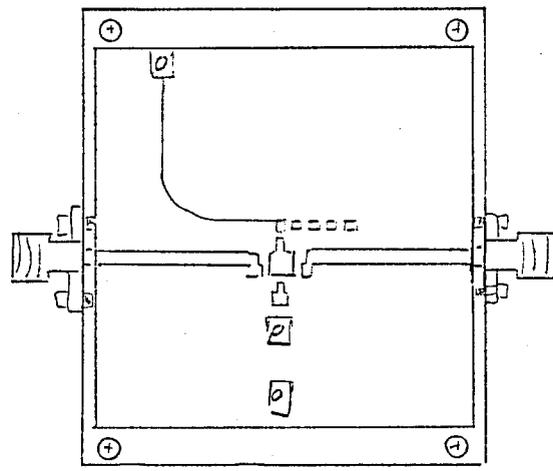


(a) Modulator/Photodetector

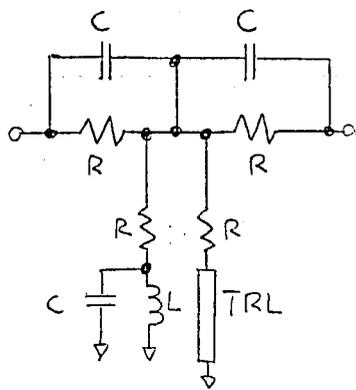
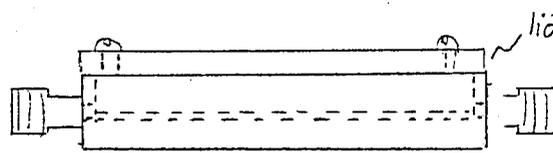


(b) Single Amplifier

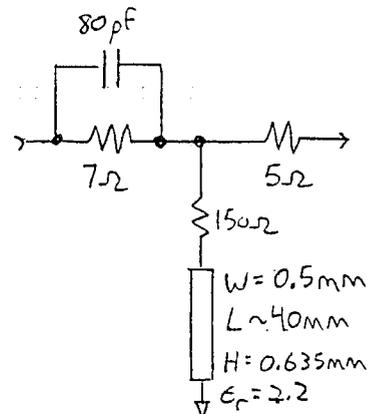
Fig. 2



Physical layout
(scale 1:1)



General purpose equalizer



Initial design

Fig. 3 (a)

CH1 RFI/M log MAG 10 dB/ REF 0 dB 1 -31.157 dB
CH2 TRN/M log MAG 1 dB/ REF -3.9 dB 1: -3995 dB

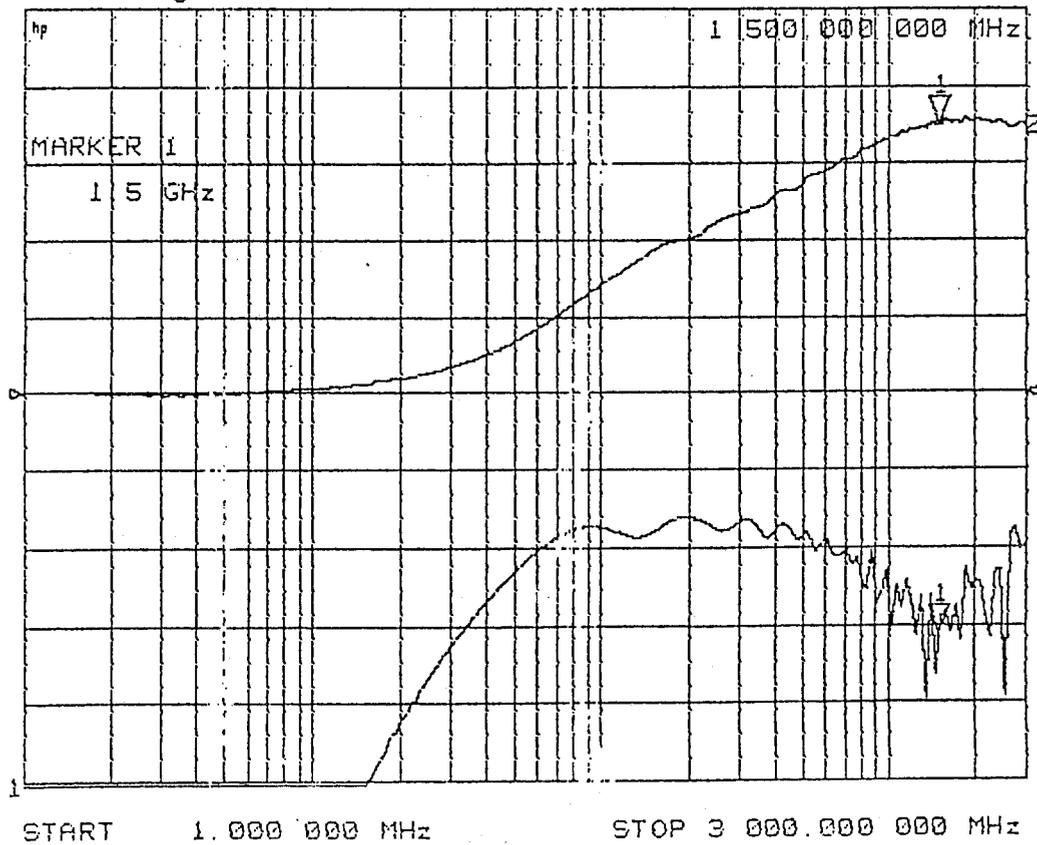


Fig. 3(b)

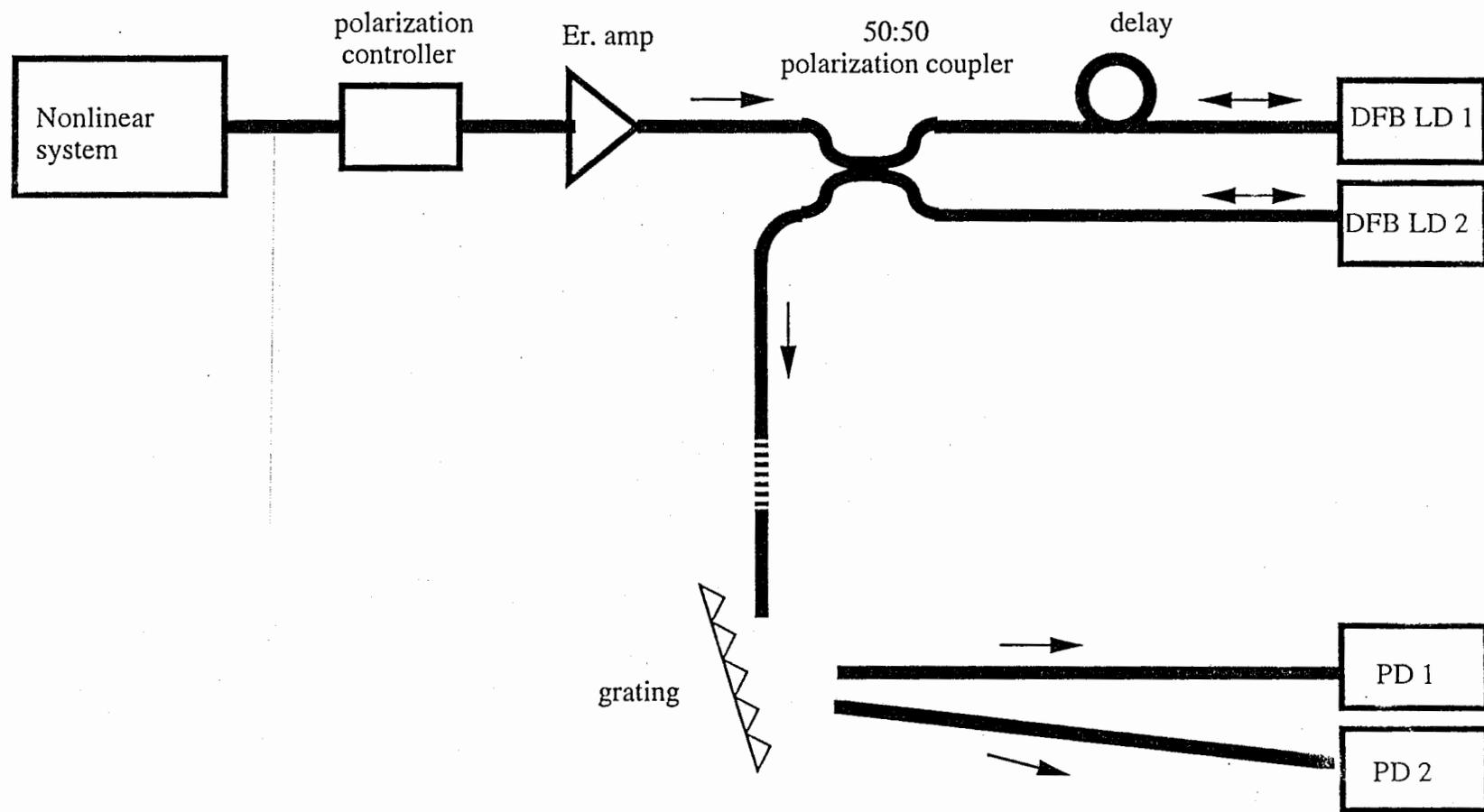


Fig. 4

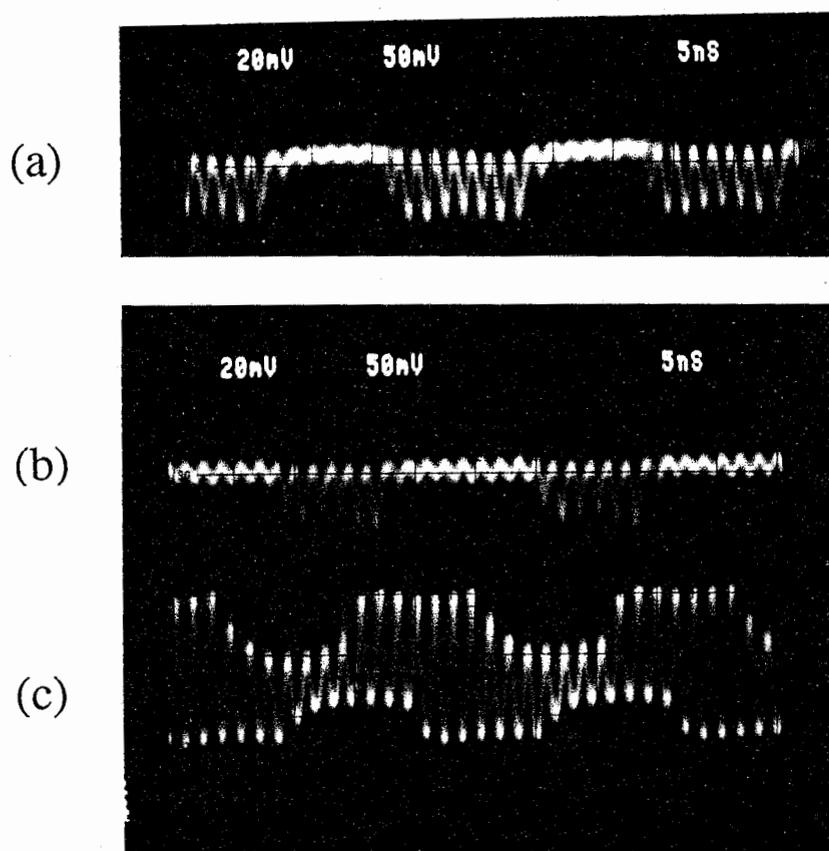


Fig. 5

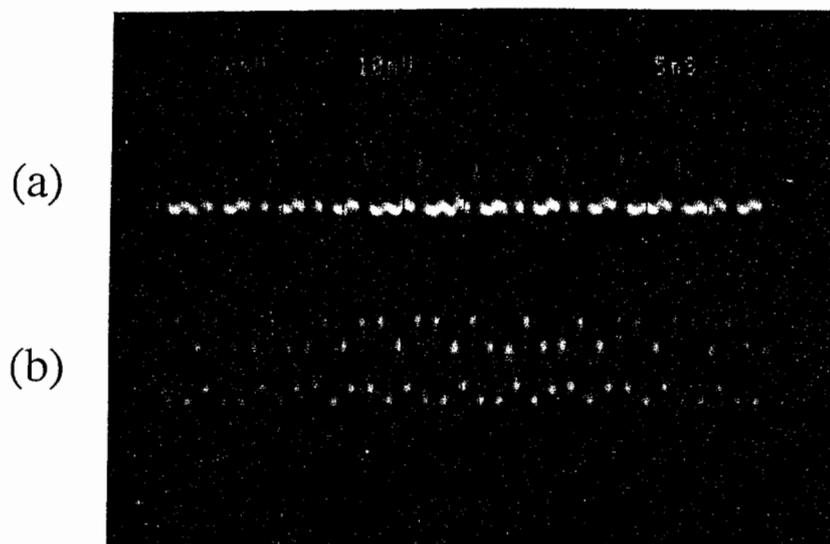


Fig. 6